

DESCRIPTION

The SG2273/3273 is a monolithic power operational amplifier, which features a high-current, low-saturation voltage, flyback protected output stage optimized for driving heavily inductive loads. Capable of operation in a single supply mode from as low as 4.5V up to 13.2V, the SG2273/3273 is ideally suited for the computer peripheral environment, driving small motors, solenoids, and linear actuators in an H-bridge configuration.

As a general-purpose op amp, the

SG2273/3273 exhibits low input offset voltage, high open loop gain, low quiescent current, a large differential input voltage range, and a common-mode input voltage range, which includes ground (V_{EE}).

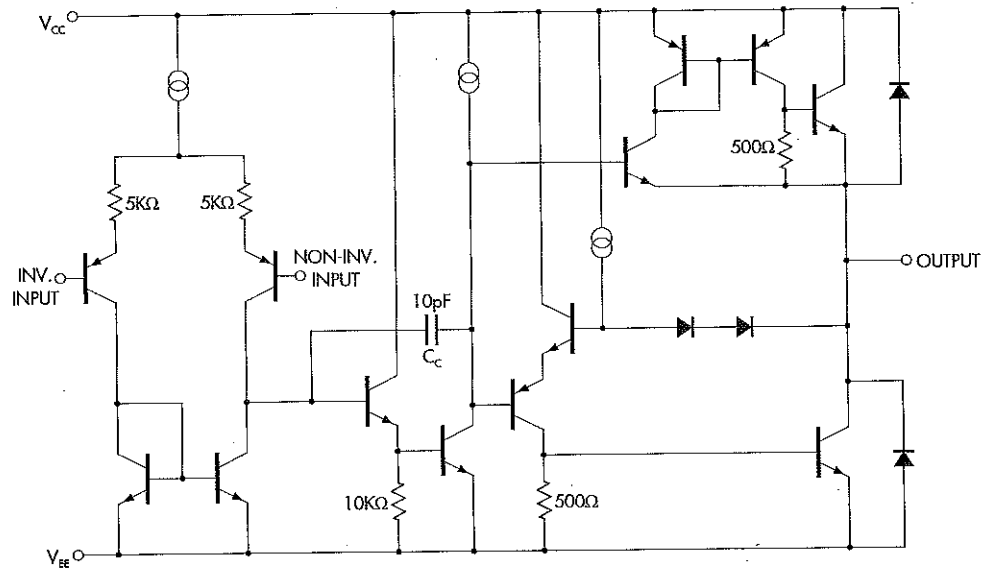
Available in a 5-pin TO-220 package, the SG2273/3273 provides system designers with a low-cost, convenient way to minimize power dissipation and reduce board area consumption in applications requiring high-current inductive load capability.

KEY FEATURES

- FULL OUTPUT SWING AT $\pm 1A$
- HIGH INDUCTIVE LOAD DRIVE CAPABILITY
- INTERNAL FLYBACK PROTECTION DIODES
- LOW POWER DISSIPATION
- SINGLE OR SPLIT SUPPLY OPERATION
- COMMON-MODE RANGE INCLUDES GROUND (V_{EE})
- HIGH OPEN LOOP GAIN
- LOW INPUT OFFSET VOLTAGE
- LARGE DIFFERENTIAL INPUT VOLTAGE RANGE
- THERMAL SHUTDOWN PROTECTION

PRODUCT HIGHLIGHT

SG2273 CIRCUIT SCHEMATIC DIAGRAM



PACKAGE ORDER INFO

T_A (°C)	P
0 to 70	Plastic TO-220 5-pin SG3273P
-45 to 85	SG2273P

FOR FURTHER INFORMATION CALL (714) 898-8121

11861 WESTERN AVENUE, GARDEN GROVE, CA 92841

SG2273/SG3373

POWER OPERATIONAL AMPLIFIER

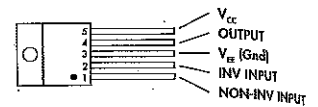
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ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage (Single Supply) (V_{CC})	-0.3V to 14V
DC Output Current (I_{OUT})	$\pm 1.4A$
Peak Output Current (Non-Repetitive) (I_{OUT})	$\pm 1.5A$
Common-Mode Input Voltage (V_{CM})	-0.3V to $V_{CC}-2V$
Differential-Mode Input Voltage (V_{DM})	$\pm V_{CC}$
Operating Junction Temperature	
Plastic (P - Package)	150°C
Storage Temperature Range	-65°C to 150°C
Lead Temperature (Soldering, 10 seconds)	300°C

Note 1. Values beyond which damage may occur. All voltages are specified with respect to ground, and all currents are positive into the specified terminal.

PACKAGE PIN OUTS



P PACKAGE
(Top View)

THERMAL DATA

P PACKAGE:

THERMAL RESISTANCE-JUNCTION TO CASE, θ_{JC}	4.0°C/W
THERMAL RESISTANCE-JUNCTION TO AMBIENT, θ_{JA}	55°C/W

Junction Temperature Calculation: $T_j = T_A + (P_D \times \theta_{JA})$.
The θ_{JA} numbers are guidelines for the thermal performance of the device/pc-board system.
All of the above assume no ambient airflow.

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RECOMMENDED OPERATING CONDITIONS (Note 2)

Parameter	Symbol	Recommended Operating Conditions			Units
		Min.	Typ.	Max.	
Supply Voltage (Single Supply)	V_{CC}	4.5		13.2	V
Output Current	I_{OHT}		± 1.2		A
Common-Mode Input Voltage	V_{ICM}	0		$V_{CC}/2$	V
Differential-Mode Input Voltage	V_{IDM}		$\pm V_{CC}$		V
Operating Ambient Temperature Range:					
SG2273	T_A	-40		85	$^{\circ}C$
SG3373	T_A	0		70	$^{\circ}C$

Range over which the device is guaranteed functional.

ELECTRICAL CHARACTERISTICS

Unless otherwise specified, these specifications apply over the operating ambient temperatures of $-40^{\circ}C \leq T_A \leq 85^{\circ}C$ for the SG2373 and $0^{\circ}C \leq T_A \leq 70^{\circ}C$ for the SG3373; $V_{CC}=12V$. Low duty cycle pulse testing techniques are used which maintains junction and case temperatures equal to the ambient temperature.)

Parameter	Symbol	Test Conditions	SG2273			SG3273			Units	
			Min.	Typ.	Max.	Min.	Typ.	Max.		
DC Characteristics										
Offset Voltage	V_{IO}	$T_A = 25^{\circ}C$	-15	0	15	-15	0	15	mV	
			-40		40	-30		30	mV	
Bias Current	I_b	$T_A = 25^{\circ}C$	-1.0	-0.2		-1.0	-0.2		μA	
Offset Current	I_{OS}	$T_A = 25^{\circ}C$	-50		50	-50		50	nA	
			-200		200	-200		200	nA	
Differential Input Resistance	R_{ID}		500		500				$K\Omega$	
Positive Side Output Saturation Voltage	$+V_{SAT}$	$I_{OHT} = 100mA$		0.8	1.0		0.8			V
		$I_{OHT} = 500mA$		1.0	1.5		1.0	1.5		V
		$I_{OHT} = 1A$		1.4	2.0		1.4	2.0		V
Negative Side Output Saturation Voltage	$-V_{SAT}$	$I_{OHT} = 100mA$		0.3	0.7		0.3			V
		$I_{OHT} = 500mA$		0.6	1.0		0.6	1.0		V
		$I_{OHT} = 1A$		1.3	2.0		1.3	2.0		V
Open Loop Voltage Gain	A_{VOL}		70	90		70	90		dB	
Common-Mode Rejection Ratio	CMRR	$T_A = 25^{\circ}C$	66	90		66	90		dB	
Power Supply Rejection Ratio	PSRR		60	80		60	80		dB	
Quiescent Drain Current	I_{CC}	$T_A = 25^{\circ}C$		7	17		7	15		mA
Thermal Shutdown Temperature		$T_A = 25^{\circ}C$					175			$^{\circ}C$
AC Characteristics ($T_A = 25^{\circ}C$)										
Gain Bandwidth Product	GBWP	$R_L = \infty\Omega$		800			800			KHz
Rate of Change	dV_O/dt	$AV = 1$		1.6			1.6			V/ μs
Bandwidth, -3dB	PBW			200			200			KHz
Input Noise Voltage	E_N	22Hz to 22KHz		10			10			μV
Input Noise Current	I_N	22Hz to 22KHz		200			200			pA
Common Mode Separation	CS	$f = 1KHz, R_L = 10\Omega, AV_{CL} = 30dB$		60			60			dB



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Characteristic Curves

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CHARACTERISTIC CURVES

FIGURE 1. — LARGE SIGNAL TRANSIENT RESPONSE

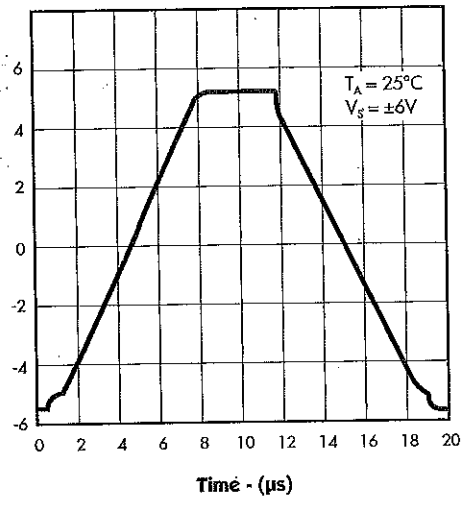


FIGURE 2. — SMALL SIGNAL TRANSIENT RESPONSE

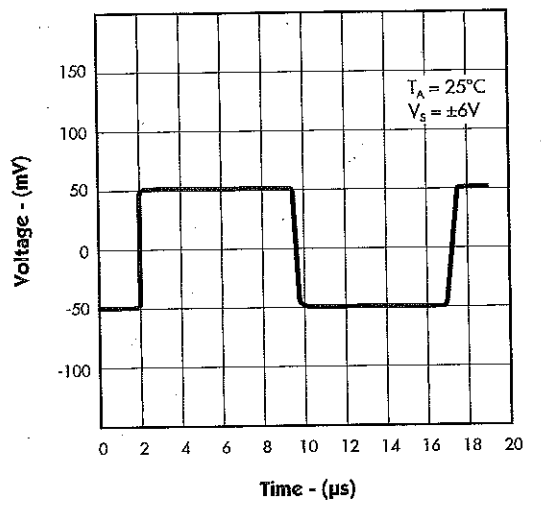


FIGURE 3. — COMMON-MODE REJECTION RATIO vs. FREQUENCY

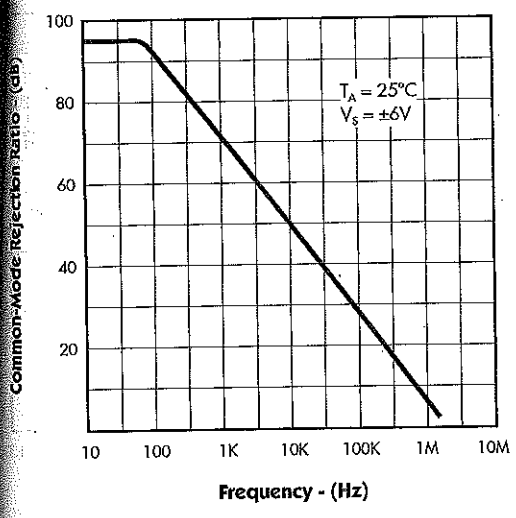
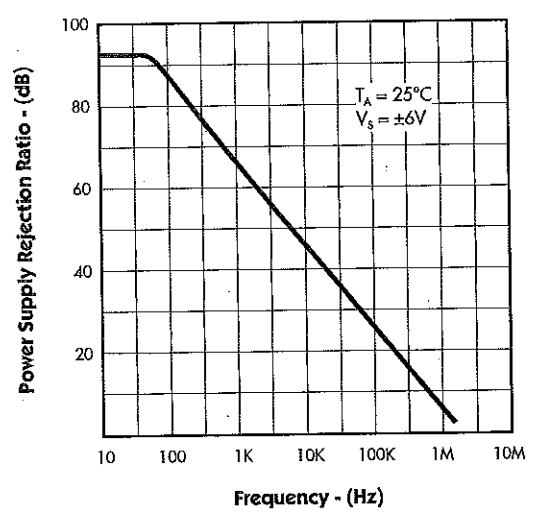


FIGURE 4. — POWER SUPPLY REJECTION vs. FREQUENCY



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CHARACTERISTIC CURVES

FIGURE 5. — OPEN LOOP GAIN vs. FREQUENCY

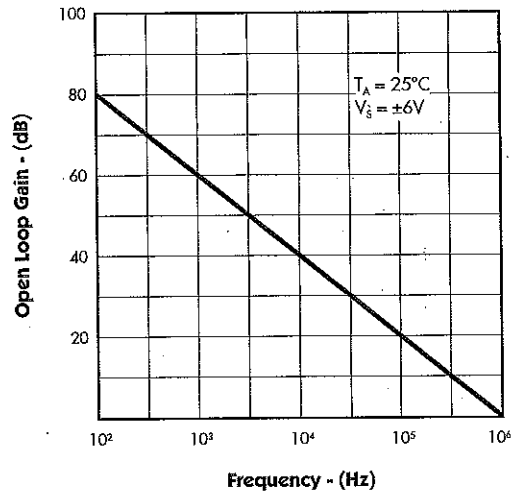


FIGURE 6. — SUPPLY CURRENT vs. SUPPLY VOLTAGE

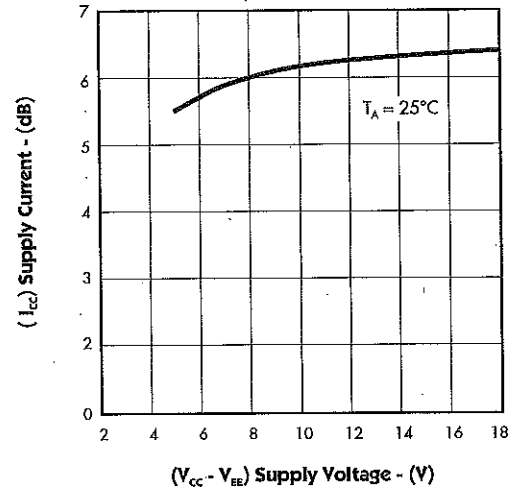


FIGURE 7. — SUPPLY CURRENT vs. TEMPERATURE

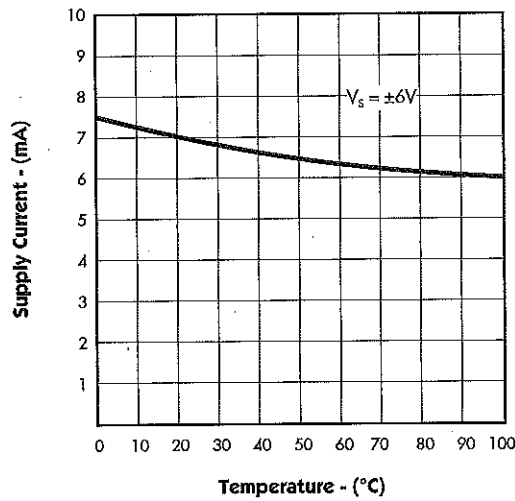
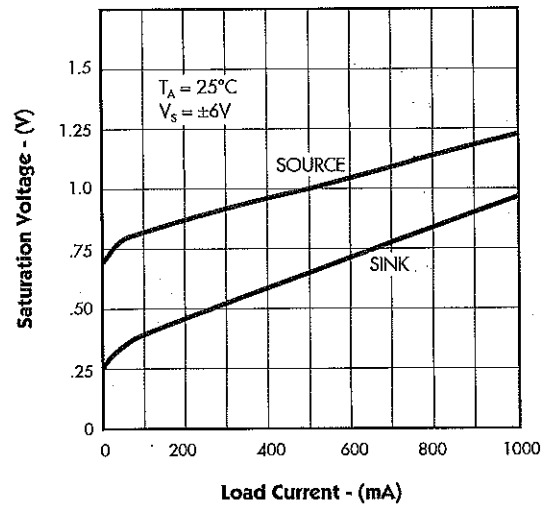


FIGURE 8. — SATURATION VOLTAGE vs. LOAD CURRENT

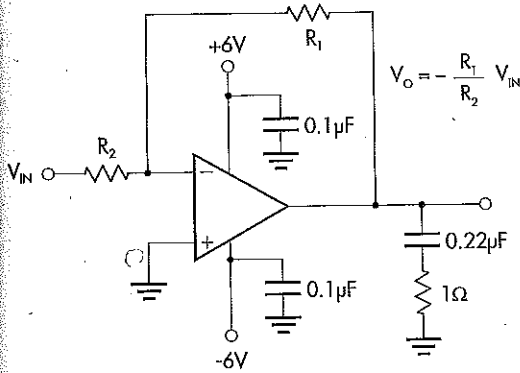


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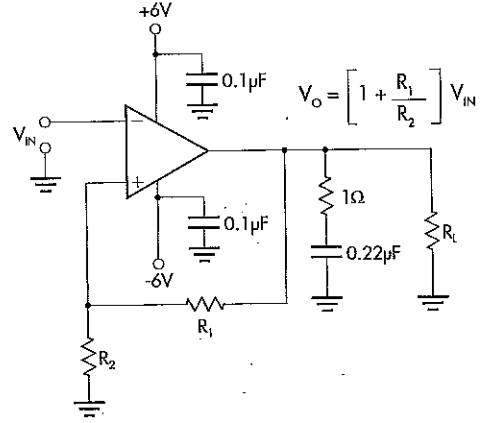
TYPICAL APPLICATION CIRCUITS

FIGURE 9. — INVERTING POWER AMPLIFIER



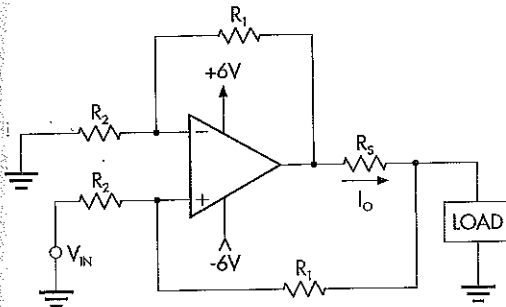
$$V_o = -\frac{R_1}{R_2} V_{IN}$$

FIGURE 10. — NON-INVERTING POWER AMPLIFIER



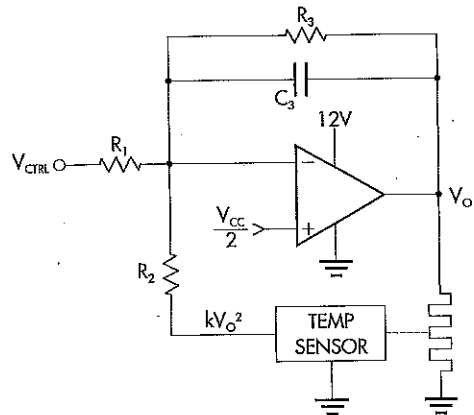
$$V_o = \left[1 + \frac{R_1}{R_2} \right] V_{IN}$$

FIGURE 11. — REGULATED CURRENT SOURCE FOR A GROUNDLED LOAD



$$I_o = \frac{R_1}{R_2 + R_s} V_{IN} \quad R_s \ll R_1 \text{ and } R_2$$

FIGURE 12. — ADJUSTABLE TEMPERATURE CONTROL



$$R_3 \gg R_1 \text{ and } R_2$$

$$V_o = \sqrt{\frac{1}{k} \left[\frac{V_{CC}}{2} \left(1 + \frac{R_1}{R_2} \right) - \frac{R_2}{R_1} V_{CTRL} \right]}$$

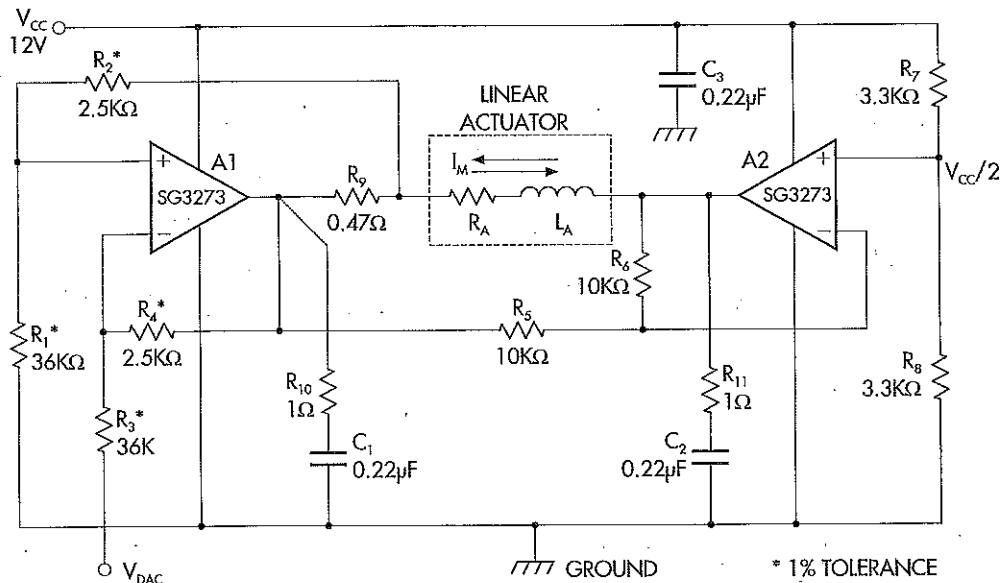
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TYPICAL APPLICATION CIRCUITS

FIGURE 13. — 3.5-INCH WINCHESTER DISK DRIVE HEAD POSITION CONTROL AMPLIFIER



if $\frac{R_1}{R_2} = \frac{R_3}{R_4}$

then $|I_M| = \frac{R_4}{R_3 R_A} V_{DAC} \begin{cases} I_M: \text{Sink to } A_2 \text{ op-amp if } V_{DAC} < 0 \\ I_M: \text{Sink to } A_1 \text{ op-amp if } V_{DAC} > 0 \end{cases}$

For the opposite example:

$I_M = \frac{2.5K}{(36K)(0.47)} V_{DAC} = 0.148 V_{DAC} \begin{cases} |I_M| \leq 740\text{mA if } \\ -5V \leq V_{DAC} \leq 5V \end{cases}$