

## 2.7V to 5.5V, 6A Constant Frequency Hysteretic Synchronous Buck Regulator with I<sup>2</sup>C Production Data Spec



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### Description

The LX7220 is a digitally controlled step-down regulator IC with an integrated 22mΩ high-side P-channel MOSFET and a 13mΩ low-side N-channel MOSFET. It features Microsemi's proprietary constant-frequency hysteretic control engine for near-instantaneous correction to line/load transients. It does not require high-ESR output capacitors and incorporates energy-saving "PSM" (Power Save or Pulse Skip Mode) at light loads, to extend battery life in mobile applications.

The LX7220 has an I<sup>2</sup>C serial interface port for output voltage margining and monitoring if required (it can also operate in default mode). In addition, it includes robust fault monitoring functions.

The LX7220 will operate from 2.7V to 5.5V, and is available in 0.95V or 0.9V output voltages (no voltage divider is necessary). The output voltage can also be adjusted with an input voltage of 5V and external voltage divider up to 3.3V.

### Features

- ◆ Constant Frequency Hysteretic Control
- ◆ Extremely Fast Line/Load Transient Response
- ◆ I<sup>2</sup>C for Output Adjustment (3.4Mbps)
- ◆ 1.2 MHz Switching Frequency
- ◆ Extremely Low-R<sub>DS(ON)</sub> MOSFETS
- ◆ Input Voltage Rail 2.7V to 5.5V
- ◆ Greater than 6A Output Current
- ◆ Default Power Save Mode for Light-Load Efficiency
- ◆ UVLO, OVP, OCP
- ◆ -40°C to +85°C Ambient Temperature
- ◆ Available in VQFN 2mm x 3mm 14L Package
- ◆ RoHS Compliant

### Applications

- High Performance HDD
- Solid-State Drive
- Data Center Applications
- Raid/Host Bus Adaptors
- Optical Transceivers

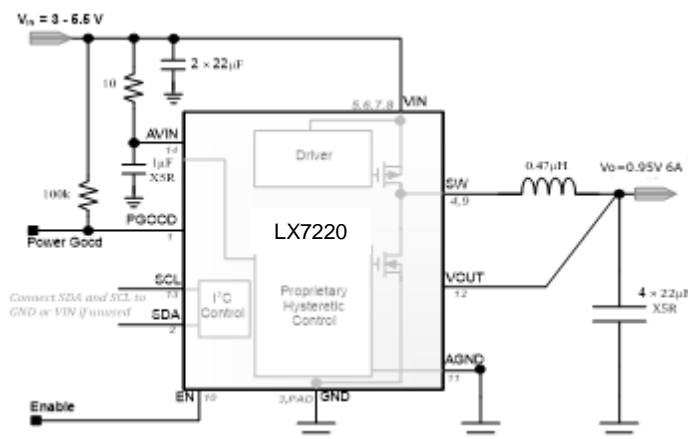


Figure 1: Typical 5V to 0.95V at 6A schematic



### Pin/Ball Configuration

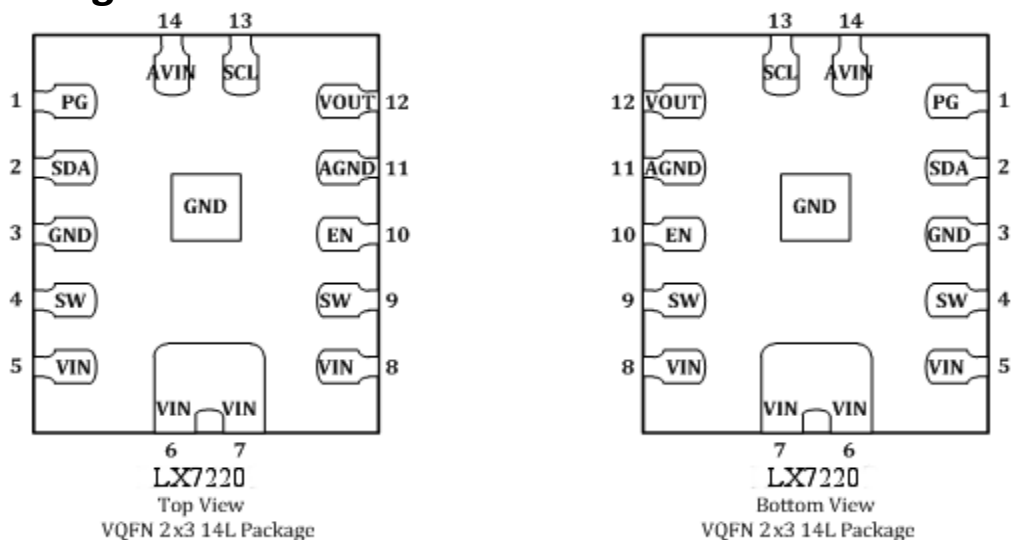


Figure 4: Pinout

### Ordering Information

Temperature	Package Type	Part Marking	Part Number	y-Output Voltage	x-Slave Address A1A0**	Packaging Type
-40°C to 85°C	VQFN 2x3 14L RoHS compliant, Pb-free	MSCN 7220	LX7220-03ILQ-TR	3=0.95V	0=E0h	Tape and Reel
		MSCP 7220	LX7220-13ILQ-TR		1=E2h	
		MSCR 7220	LX7220-23ILQ-TR		2=E4h	
		MSCS 7220	LX7220-33ILQ-TR		3=E6h	
		MSCJ 7220	LX7220-02ILQ-TR	2=0.9V	0=E0h	
		MSCK 7220	LX7220-12ILQ-TR		1=E2h	
		MSCL 7220	LX7220-22ILQ-TR		2=E4h	
		MSCM 7220	LX7220-32ILQ-TR		3=E6h	

\* Consult factory for other I<sup>2</sup>C slave address and set output voltage options. (LX7220-xyILQ-TR)

"x" is the 2 LSB bits of the binary I<sup>2</sup>C slave address (0 to 3);

"y" is the set output voltage (3 is 0.95V, 2 is 0.9V, 1 is 0.8, 0 is 0.85)

\*\*Refer to Page 12 Table 1: I2C Slave Address

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### Pin/Ball Description

Pin/Ball Number	Pin/Ball Designator	Description
1	PGOOD	Open Drain status output, requires external pull up resistor. This pin will go low when VOUT is outside the defined power good range, when the die is hotter than the thermal shutdown threshold, when PVIN is above the over voltage threshold, or when PVIN is below the under voltage threshold. PGOOD will go high after the last of these fault conditions clear.
10	EN	Enable for switching regulator. Force high to enable, force low to disable the IC.
13	SCL	Serial clock input for I <sup>2</sup> C. Connect directly to GND if unused.
12	VOUT	Output voltage sense. Connect directly to output rail or resistive voltage divider output.
2	SDA	Serial data bus (bidirectional) for I <sup>2</sup> C. Connect directly to GND if unused.
3, PAD	GND	Ground. Connect to ground plane.
11	AGND	Analog Ground. Connect to ground plane.
5,6,7,8	VIN	Input of IC and buck stage. Connect to input rail VIN (between 2.7V and 5.5V). A minimum input capacitance of one 1 $\mu$ F and one 22 $\mu$ F of X5R or better multilayer ceramic, should be placed very close to IC between this node and GND.
14	AVIN	Analog VIN voltage input pin.
4,9	SW	Switching Node. Drives the external L-C low pass filter.



Functional Block Diagram

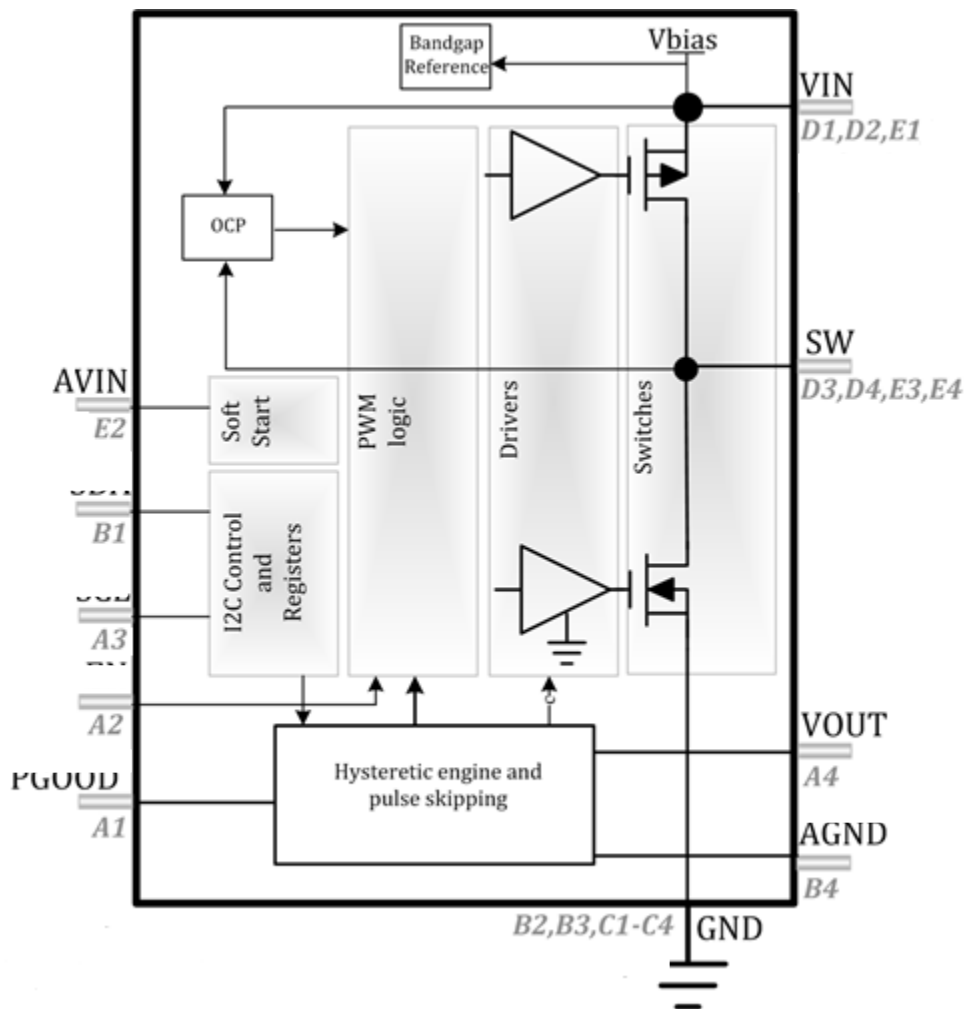


Figure 5: Block Diagram

## 2.7V to 5.5V, 6A Constant Frequency Hysteretic Synchronous Buck Regulator with I<sup>2</sup>C

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### Absolute Maximum Ratings

Performance is not necessarily guaranteed over this entire range. These are maximum stress ratings only. Exceeding these ratings, even momentarily, can cause immediate damage, or negatively impact long-term operating reliability.

	Min	Max	Units
VIN, SW to GND	-0.3	7	V
AVIN,VOUT, SDA, SCL, EN, PGOOD to GND	-0.3	7	V
SW to GND (Shorter than 50ns)	-2	7	V
Maximum Junction Temperature		150	°C
Lead Soldering Temperature (30s, reflow)		260 (+0, -5)	°C
Storage Temperature	-65	150	°C

### Operating Ratings

Performance is generally guaranteed over this range as further detailed below under Electrical Characteristics.

	Min	Max	Units
VIN	2.7	5.5	V
Ambient Temperature	-40	85	°C
Output Current	0	6	A

**Note:** Corresponding Absolute Max Junction Temperature is 150°C.

### Thermal Properties

Thermal Resistance	Typ	Units
$\theta_{JA}$ VQFN	50	°C/W

**Note:** The  $\theta_{JA}$  numbers assume no forced airflow. Junction Temperature is calculated using  $T_J = T_A + (P_D \times \theta_{JA})$ . In particular,  $\theta_{JA}$  is a function of the PCB construction. The stated number above is for a four-layer board in accordance with JESD-51 (JEDEC).

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## Electrical Characteristics

The following specifications apply over the operating ambient temperature of  $-40^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$  except where otherwise noted with the following test conditions:  $V_{IN} = 5\text{V}$ ,  $EN = 5\text{V}$ ,  $SCL = 5\text{V}$ ,  $SDA = 5\text{V}$ , default register settings. Typical values stated, are either by design or by production testing at  $25^{\circ}\text{C}$  ambient.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
<b>Input Voltage</b>						
$I_Q$	Input current	$I_{LOAD} = 0$ , PSM enabled	200	440	600	$\mu\text{A}$
$I_{IN}$	Input current at shut down	$EN = \text{GND}$ , $T_A = 25^{\circ}\text{C}$		0.1	14	$\mu\text{A}$
$I_{IN\_I^2C}$	Input current I <sup>2</sup> C shut down	$V_{SEL}(7) = \text{low}$ , $EN = \text{high}$		100	120	$\mu\text{A}$
$UVLO$	Under voltage rising threshold	$V_{IN}$ rising		2.6	2.89	V
$UVLO_{HYST}$	UVLO hysteresis			0.26		V
$OVP_R$	Over voltage rising threshold		6.0		6.30	V
$OVP_F$	Over voltage falling hysteresis			0.2		V
<b>Reference Voltage</b>						
$T_{SS}$	$V_{REF}$ slew rate	$SLEW: \text{Ctrl2}(2:1) = 01$		0.8		$\text{mV}/\mu\text{s}$
$T_{HICCUP}$	Hiccup time	$V_{OUT} = 0.2\text{V}$		9.8		ms
<b>Output Voltage</b>						
$V_{OUT}$	Default $V_{OUT}$	$V_{OUT}=0.95\text{V}$ ( $V_{in}=2.7\text{V} - 5\text{V}$ ), $V_{SEL}=40\text{h}$	0.94	0.95	0.959	V
	$V_{OUT}$ I2C $V_{SEL}$	$V_{OUT}=0.75\text{V}$ ( $V_{in}=2.7\text{V} - 5\text{V}$ ), $V_{SEL}=20\text{h}$	0.743	0.75	0.758	V
		$V_{OUT}=0.75\text{V}$ ( $V_{in}=2.7\text{V} - 5\text{V}$ ), $V_{SEL}=20\text{h}$ $-10^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$	0.741	0.75	0.759	V
		$V_{OUT}=1.197\text{V}$ ( $V_{in}=2.7\text{V} - 5\text{V}$ ), $V_{SEL}=7\text{Fh}$	1.185	1.197	1.2	V
	Line regulation	$V_{IN}$ from 3V to 5.5V, $I_{LOAD} = 1\text{A}$ . Note 1		0.1		%
	Load regulation	$I_{LOAD} = 0\text{A}$ to 5A. Note 1		-0.23		%/A
	$V_{OUT}$ input current			0	1	$\mu\text{A}$
$V_{ouv}$	$V_{OUT}$ under voltage threshold	$V_{OUT}$ below this threshold will initiate a hiccup sequence	77	82	85	$\%V_{REF}$
<b>SW</b>						

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Symbol	Parameter	Conditions	Min	Typ	Max	Units
R <sub>DS(on)_H</sub>	High side on resistance	V <sub>IN</sub> = 5V		22		mΩ
R <sub>DS(on)_L</sub>	Low side on resistance	V <sub>IN</sub> = 5V		13		mΩ
OCP	Current limit	Note 1	7.5	8.5	10.0	A
T <sub>SH</sub>	Thermal shut down threshold	Note 1		150		°C
T <sub>H</sub>	Hysteresis	Note 1		20		°C
F <sub>SW</sub>	PWM switching frequency		1.0	1.2	1.4	MHz
R <sub>SWDISC</sub>	SW discharge resistance	EN = low; Discharge: Ctrl2(4) = 1	80	200	1400	Ω
<b>EN, SDA (as input), SCL</b>						
V <sub>IH</sub>	Input high		1.1			V
V <sub>IL</sub>	Input low				0.4	V
V <sub>H</sub>	Hysteresis		0.05	0.15		V
I <sub>II</sub>	Input current			0	1.1	μA
V <sub>OL</sub>	Low level output voltage	Logic0 output voltage, I <sub>sink</sub> =2mA(Note 1)	0		0.2*V <sub>DD</sub>	V
I <sub>OL</sub>	Low level output current	V <sub>ol</sub> =0.4V(Note 1)	3			mA
<b>PGOOD</b>						
V <sub>PG90</sub>	PGOOD VOUT lower threshold	VOUT rising, percentage of V <sub>REF</sub>	82	85	88	%V <sub>REF</sub>
V <sub>PG110</sub>	PGOOD VOUT upper threshold	VOUT falling, percentage of V <sub>REF</sub>	105	110	115	%V <sub>REF</sub>
V <sub>PGHY</sub>	Hysteresis	Percentage of V <sub>REF</sub>		5		%V <sub>REF</sub>
P <sub>GRDSON</sub>	PGOOD pull down resistance			13	20	Ω
	PGOOD leakage current			0	1	μA
	PGOOD delay	PGOOD rising edge delay	27	45	69	ms
<b>7 Bit DAC</b>						
	Differential linearity	Monotonicity assured by design			0.8	LSB

Note 1: Guaranteed by design.

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### Application Specifics

<b>Efficiency</b>	I <sub>OUT</sub> = 2.0A, VCC = 5V, VOUT = 3.3V	95%
	I <sub>OUT</sub> = 4.0A, VCC = 5V, VOUT = 0.8V, Inductor(IHLP-2020CZ-01)	>84%
	I <sub>OUT</sub> = 4.0A, VCC = 5V, VOUT = 0.9V, Inductor(SPM5015)	>83%
<b>VOUT Max Transient</b>	100mA↔4A, 1A/2us, C <sub>LOAD</sub> = 4 x 22μF ceramic caps, 0.47μH inductor. Step Duration 1us-50us	Peak to peak<80mV
<b>Typical Load Inductance</b>	IHLP2020CZ(DCR = 6.7mΩ, IDC = 12.2A, ISAT = 16A) SPM5015(DCR = 16.3mΩ, IDC = 7A, ISAT = 13.8A)	0.47μH
<b>Typical Load Capacitance</b>	6.3V, X5R	4x22μF
<b>Vin, VOUT Ripple Noise Measurement (EMI)</b>	I <sub>out</sub> =5A, VCC=5V, VOUT=1V, Inductor(SPM5015)	<200mVpp <50mVpp
<b>Startup Vin inrush current with Vout pre-bias</b>	VCC=5V, R <sub>LOAD</sub> =50 Ohms, C <sub>LOAD</sub> =200uF ceramic caps, Enable F=500Hz 90% duty cycle	<175mA <sub>PEAK</sub>



## 2.7V to 5.5V, 6A Constant Frequency Hysteretic Synchronous Buck Regulator with I<sup>2</sup>C Production Data Spec



### I<sup>2</sup>C Timing Specifications

Symbol	Parameter	Conditions	C <sub>b</sub> = 100 pF (max) (*Note2)		C <sub>b</sub> = 400 pF		Unit
			Min	Max	Min	Max	
f <sub>SCHL</sub>	SCL clock frequency		0	3.4	0	0.4	MHz
t <sub>SU;STA</sub>	Set-up time for a repeated START condition		160	-	600	-	ns
t <sub>HD;STA</sub>	Hold time (repeated) START condition		160	-	600	-	ns
t <sub>LOW</sub>	LOW period of the SCL clock		160	-	1300	-	ns
t <sub>HIGH</sub>	HIGH period of the SCL clock		60	-	600	-	ns
t <sub>SU;DAT</sub>	Data set-up time		10	-	100	-	ns
t <sub>HD;DAT</sub>	Data hold time		0	70	0	-	ns
t <sub>rCL</sub>	Rise time of SCL signal		10	40	20*0.1C <sub>b</sub>	300	ns
t <sub>rCL1</sub>	Rise time of SCL signal after a repeated START condition and after an acknowledge bit		10	80	20*0.1C <sub>b</sub>	300	ns
t <sub>fCL</sub>	Fall time of SCL signal		10	40	20*0.1C <sub>b</sub>	300	ns
t <sub>rDA</sub>	Rise time of SDA signal		10	80	20*0.1C <sub>b</sub>	300	ns
t <sub>fDA</sub>	Fall time of SDA signal		10	80	20*0.01C <sub>b</sub>	300	ns
t <sub>SU;STO</sub>	Set-up time for STOP condition		160	-	600	-	ns
t <sub>BUF</sub>	Bus free time between a STOP and START condition		160	-	1300	-	ns
t <sub>VD;DAT</sub>	Data valid time		-	160	-	900	ns
t <sub>VD;ACK</sub>	Data valid acknowledge time		-	160	-	900	ns
C <sub>b</sub>	Capacitive load for each bus line	SDA and SCL lines	-	100	-	400	pF

Note 1: All values referred to V<sub>IH</sub>(min) and V<sub>IL</sub>(max) levels of I/O stages table.

Note 2: Loads in excess of 100pF will restrict bus operation speed below 3.4MHz

## 2.7V to 5.5V, 6A Constant Frequency Hysteretic Synchronous Buck Regulator with I<sup>2</sup>C Production Data Spec



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## Operation Theory

### Basic Operation

The LX7220 compares V<sub>OUT</sub> voltage to an internal reference, V<sub>REF</sub>. When V<sub>OUT</sub> is lower than V<sub>REF</sub>, the upper switch turns on and the lower switch turns off. When V<sub>OUT</sub> is higher than V<sub>REF</sub>, the upper switch turns off and the lower switch turns on. An internal ramp helps to keep the switching frequency constant over a wide range of output capacitor values and parasitic components (i.e. ESR, ESL). In addition, a frequency control loop keeps the switching frequency constant during continuous conduction mode.

At light loads, if enabled, the converter automatically reduces the switching frequency and enters discontinuous conduction to optimize efficiency while ensuring low V<sub>OUT</sub> ripple voltage.

An integrated I<sup>2</sup>C bus interface, operating up to 3.4Mbps, adds the following use programmability to the converter:

1. On the fly programming of the output voltage in 4.7mV increments.
2. Enable / Disable the regulator.
3. Allow PSM or limit operation to only PWM mode.
4. Set the V<sub>REF</sub> slew rate.
5. Switch node slew rate control.

### Setting the Output Voltage

The output voltage is set with the reference voltage and how the V<sub>OUT</sub> pin (12) is connected to the output. With a direct connection (i.e. See Figure 2), the reference voltage equals the output voltage. When the V<sub>OUT</sub> pin (12) is connected to a resistor divider (i.e. See Figure 3), this also determines the output voltage. **At startup, the reference voltage is determined by the parts number “y” parameter (i.e. LX7220-xyILQ).** “y” sets the output voltage (3 is 0.95V, 2 is 0.9V). After startup, the reference voltage can be programmed with the I<sup>2</sup>C bus VSEL register value.

$$V_{REF} = 0.6V + N_{SEL} \cdot 0.0046875V \quad (2)$$

Where N<sub>SEL</sub> is the decimal value of the 7 VSEL bits.

The output voltage is determined as follows:

$$V_{OUT} = V_{REF} \times \left( 1 + \frac{R_{TOP}}{R_{BOTTOM}} \right) \quad (3)$$

R<sub>TOP</sub> is the resistor connected from V<sub>OUT</sub> pin to output, R<sub>BOTTOM</sub> is the resistor connected from V<sub>OUT</sub> pin to GND.

### Startup

If the LX7220 is enabled, when V<sub>IN</sub> rises above the UVLO threshold, the regulator will initiate a startup sequence. The serial port registers are initialized to their default values and all internal bias voltages and currents are allowed to stabilize. V<sub>REF</sub> then ramps up from 0V to the default voltage at the default slew rate. At the end of the ramp time, PGOOD is allowed to go high after V<sub>OUT</sub> has reached the PGOOD rising threshold. During the ramp time, the LX7220 switches to PSM to allow discontinuous operation. This switchover is independent of the MODE bit setting.

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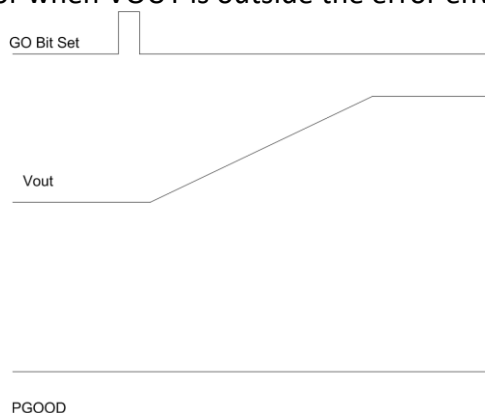
### Operation Theory (Continued)

#### Over Current Protection

The LX7220 protects against all types of short circuit conditions. Cycle by cycle over current protection turns off the upper switch when the current exceeds the OCP threshold. When this occurs, the upper switch is held off for at least 350ns before being allowed to turn on again. After startup, if V<sub>OUT</sub> drops below the V<sub>OUT</sub> under voltage threshold, a hiccup sequence will be initiated where both output switches are shut off for 6.5ms before initiating another soft start cycle. This protects against a crowbar short circuit. The V<sub>OUT</sub> under voltage detection is not active during start up.

#### Positive Voltage Transitions

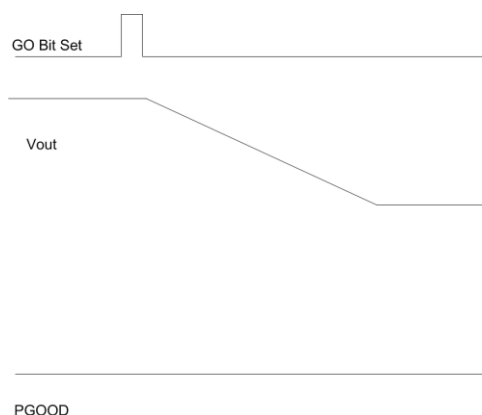
After the initial startup sequence, the output voltage can be programmed to a new value by programming the VSEL register bits and then asserting the GO bit. V<sub>REF</sub> will transition to the new value at the programmed slew rate. The PGOK monitor bit is deasserted during the V<sub>REF</sub> ramp time, or when V<sub>OUT</sub> is outside the error envelope.



**Figure 18:** Positive Voltage Transition

#### Negative Voltage Transitions

A negative voltage transition occurs when a lower output voltage is programmed into the VSEL register, and initiated by asserting the GO bit. In PSM, the LX7220 will not discharge the output filter capacitor.



**Figure 19:** Negative Voltage Transition

#### Enabling Regulator from I<sup>2</sup>C Bus

In addition to the EN pin, the regulator can be enabled and disabled via the I<sup>2</sup>C bus by programming the control register. During disable, the regulator and most of the support circuitry is turned off. However, the I<sup>2</sup>C bus circuitry is still active and may be programmed.

#### Switch Node rise rate adjustment

The LX7220 can be programmed to operate in a lower emissions mode by slowing down the switch node rise rate. In this mode, the switch node rise rate will slow down 25%, reducing the switching frequency harmonic content.

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## I<sup>2</sup>C Interface

### I<sup>2</sup>C Port Functional Description

- Simple two wire, bidirectional, serial communication port.
- Multiple devices on same bus speeds from 400Kbps (FS-Mode) to 3.4Mbps (HS-Mode).
- SOC Master controls bus.
- Device listens for the unique address that precedes data.

### General I<sup>2</sup>C Port Description

The LX7220 includes an I<sup>2</sup>C compatible serial interface, using two dedicated pins: SCL and SDA for I<sup>2</sup>C clock and data respectively. Each line is externally pulled up to a logic voltage when they are not being controlled by a device on the bus. The LX7220 interface acts as an I<sup>2</sup>C slave that is clocked by the incoming SCL clock. The LX7220 I<sup>2</sup>C port will support both the Fast mode (400kHz max) and typically the High Speed mode (3.4MHz max). The data on the SDA line must be stable during the HIGH period of the clock signal (SCL). The state of the SDA line can only be changed when SCL is LOW (except for start, stop, and restart).

### Register Map

The LX7220 has five 8-bit user-accessible registers. See Control Register Bit Definition.

### Slave Address

In the table below, the A1 and A0 are the binary value of the address given in the ordering information for the “x” value shown on page 3.

7	6	5	4	3	2	1	0
1	1	1	0	0	A1	A0	R/W

Table 1: I<sup>2</sup>C Slave Address

### START and STOP Commands

When the bus is idle, both SCL and SDA must be high except in the power up case where they may be held high or low during the system power up sequence.

The STX SOC (bus master) signals START and STOP bits signify the beginning and the end of the I<sup>2</sup>C transfer. The START condition is defined as the SDA signal transitioning from HIGH to LOW while the SCL line is HIGH. The STOP condition is defined as the SDA transitioning from LOW to HIGH while the SCL is HIGH. The STX SOC acts as the I<sup>2</sup>C master and always generates the START and STOP bits. The I<sup>2</sup>C bus is considered to be busy after START condition and free after STOP condition. During data transfer, STX SOC master can generate repeated START conditions. The START and the repeated START conditions are functionally equivalent.

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## I<sup>2</sup>C Interface (Continued)

### Data Transfers

Data is transferred in 8 bit bytes by SDA with the MSB transferred first. Each byte of data has to be followed by an acknowledge (ACK) bit. The acknowledged related clock pulse is generated by the master. The acknowledge occurs when the transmitter master releases the SDA line to a high state during the acknowledge clock. The SDA line must be pulled down by the receiver slave during the 9th clock pulse to signify acknowledgment. A receiver slave which has been addressed must generate an acknowledgement (“ACK”) after each byte has been received.

After the START condition, the STX SOC (I<sup>2</sup>C) master sends a chip address. The standard I<sup>2</sup>C address is seven bits long. Making the eighth bit a data direction bit (R/W). For the eighth bit (LSB), a “0” indicates a WRITE and a “1” indicates a READ. (For clarification, communications are broken up into 9-bit segments, one byte followed by one bit for acknowledging.) The second byte selects the register to which the data will be written. The third byte contains data to write to the selected register.

When a receiver slave doesn’t acknowledge the slave address, the data line must be left HIGH by the slave. The master can then generate a STOP command to abort the transfer. If a slave receiver does acknowledge the slave address but, sometime later in the transfer cannot receive any more data bytes, the master must again abort the transfer. This is indicated by the slave generating the not acknowledge on the first byte to follow.

The slave leaves the data line HIGH and the master generates the STOP command. The data line is also left high by the slave and master after a slave has transmitted a byte of data to the master in a read operation, but this is a not acknowledge that indicates that the data transfer is successful.

### Data Transfer Timing for Write Commands

In order to help assure that bad data is not written into the part, data from a write command is only stored after a valid STOP command has been performed.

### I<sup>2</sup>C Electrical Characteristics

The minimum HIGH and LOW periods of the SCL clock specified the I<sup>2</sup>C Timing Specification table determine the maximum bit transfer rates of, 400 kbits/s for Fast-mode devices, and 3.4 Mbits/s for HS-mode Plus. Devices must be able to follow transfers at their own maximum bit rates, either by being able to transmit or receive at that speed or by applying the I<sup>2</sup>C clock synchronization procedure, which will force the master into a wait state and stretch the LOW period of the SCL signal. Of course, in the latter case the bit transfer rate is reduced.

Figures 22 and Figure 23 show all timing parameters for the HS & FS-mode timing. The ‘normal’ START condition S does not exist in HS-mode. Timing parameters for Address bits, R/W bit, Acknowledge bit and DATA bits are all the same. Only the rising edge of the first SCL clock signal after an acknowledge bit has a larger value because the external Rp has to pull-up SCL without the help of the internal current-source.

## 2.7V to 5.5V, 6A Constant Frequency Hysteretic Synchronous Buck Regulator with I<sup>2</sup>C Production Data Spec



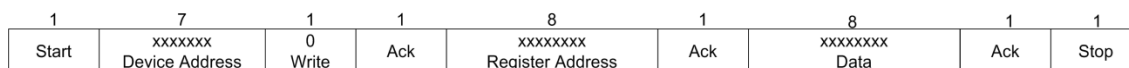
**Microsemi**

### I<sup>2</sup>C Interface (Continued)

The HS & FS-mode timing parameters for the bus lines are specified in the I<sup>2</sup>C Timing Specification Table. The minimum HIGH and LOW periods and the maximum rise and fall times of the SCL clock signal determine the highest bit rate.

With an internally generated SCL signal with LOW and HIGH level periods of 200ns and 100ns respectively, an HS-mode master fulfills the timing requirements for the external SCL clock pulses (taking the rise and fall times into account) for the maximum bit rate of 3.4 Mbit/s. So a basic frequency of 10 MHz, or a multiple of 10 MHz, can be used by an HS-mode master to generate the SCL signal. There are no limits for maximum HIGH and LOW periods of the SCL clock, and there is no limit for a lowest bit rate.

Timing parameters are independent for capacitive load up to 100 pF for each bus line allowing the maximum possible bit rate of 3.4 Mbit/s. At a higher capacitive load on the bus lines, the bit rate decreases gradually. The timing parameters for a capacitive bus load of 400 pF are specified in I<sup>2</sup>C Timing Specification Table, allowing a maximum bit rate of 1.7 Mbit/s. For capacitive bus loads between 100pF and 400 pF, the timing parameters must be interpolated linearly. Rise and fall times are in accordance with the maximum propagation time of the transmission lines SDA and SCL to prevent reflections of the open ends.



**Figure 20: Write Protocol**

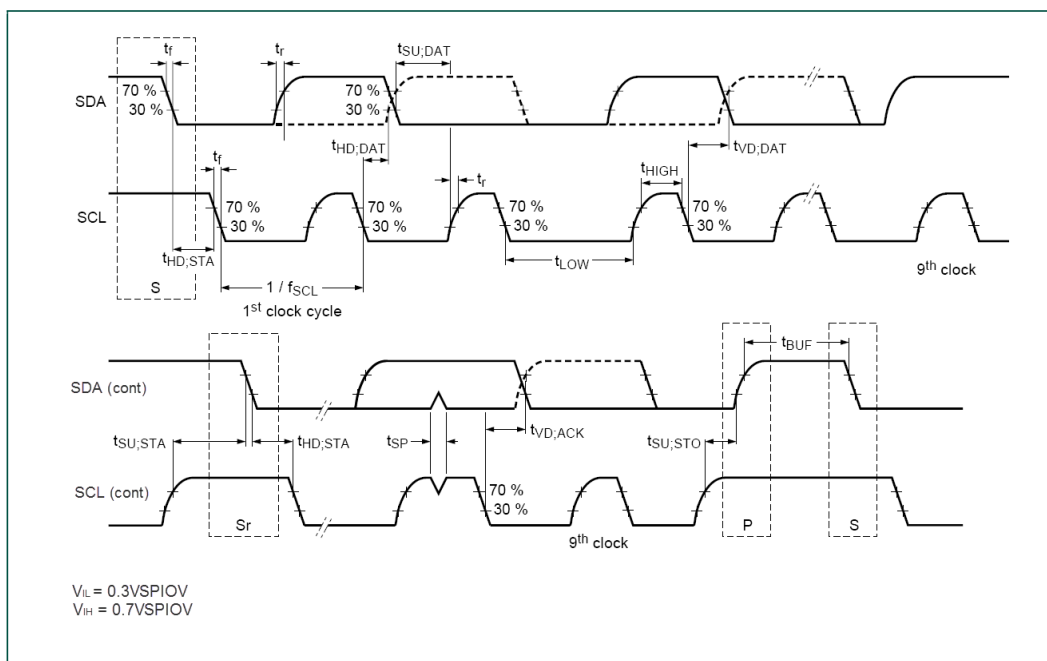


**Figure 21: Read Protocol**

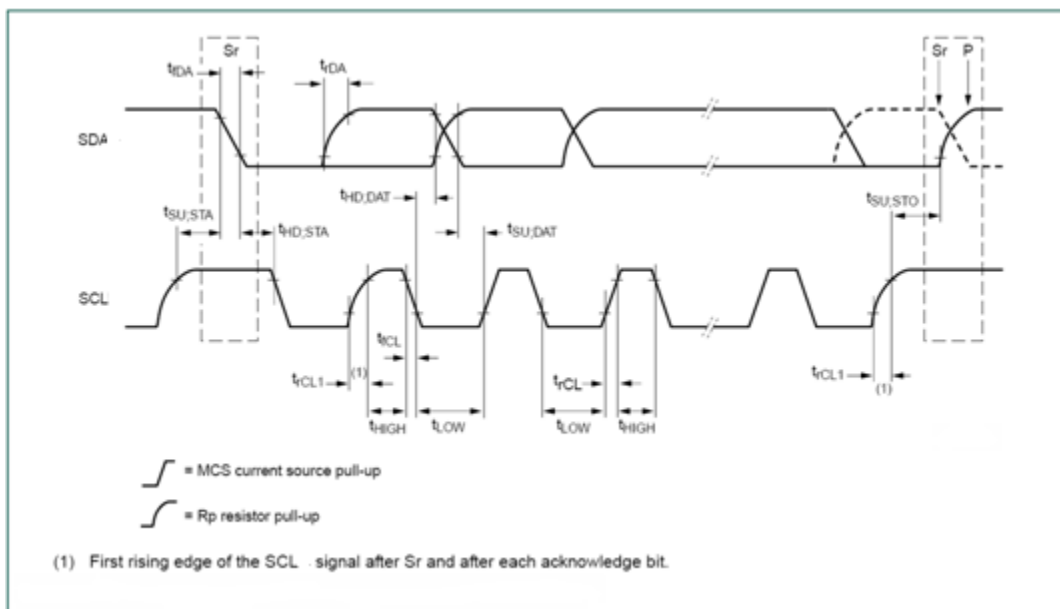


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### I<sup>2</sup>C Interface (Continued)



**Figure 22:** Definition for FS-Mode devices on the I<sup>2</sup>C Port

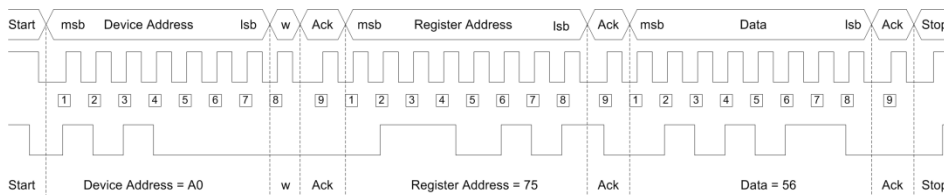


**Figure 23:** Timing definition for HS-mode devices on the I<sup>2</sup>C Port

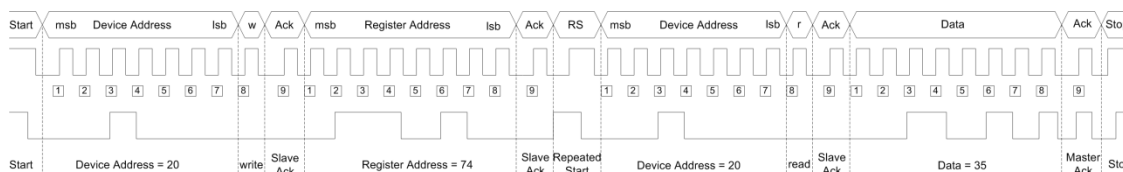


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### I<sup>2</sup>C Interface (Continued)



**Figure 24: Write Cycle Diagram**



**Figure 25: Read Cycle Diagram**

### Control Register Bit Definition

Bit	Name	Value	Description
<b>Status, Address 00h</b>			
7:3	Reserved		
2	OCP		Latched to 1 if the over current limit is reached. Write a "1" to reset the status flag.
1	OTP		Latched to 1 if an over temperature event occurs. Write a "1" to reset the status flag.
0	FB_UVLO		Latched to 1 if a FB_UVLO event occurs. Write a "1" to reset the status flag.
<b>Vsel, Address 01h, (aka dac)</b>			
7	EN	1-d	Device enabled.
		0	Device disabled.
6:0	VSEL[6:0]		7-bit DAC value to set V <sub>REF</sub> . The default value is determined by the part ordering code.
<b>Ctrl1, Address 02h, (aka reg2)</b>			
7:6	Reserved	00-d	



## 2.7V to 5.5V, 6A Constant Frequency Hysteretic Synchronous Buck Regulator with I<sup>2</sup>C Production Data Spec

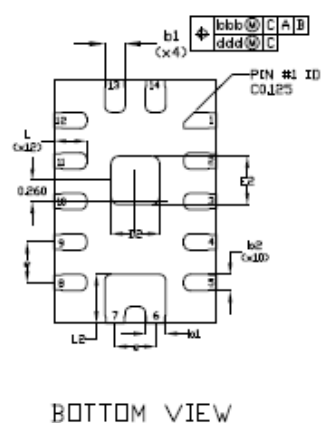
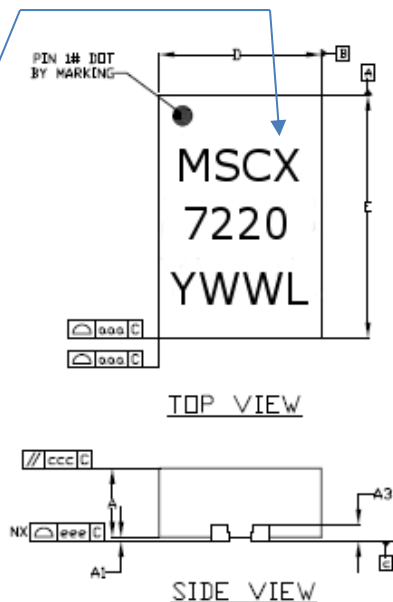


# Microsemi

Bit	Name	Value	Description
5	ctrl1	1-d	TBD
4	DLY_DIS	1	45ms delay on PGOOD is enabled.
		0-d	Disable 45ms delay on PGOOD.
3	Reserved	1-d	
		0	
2	Reserved	1-d	
		0	
1	Reserved	1-d	
		0	
0	MODE	0	PWM mode only – NO PSM.
		1-d	Power Saving Mode – allows discontinuous conduction.
<b>Vendor ID, Address 03h (Read Only)</b>			
7:4	VID[3:0]	0010	Microsemi Vendor ID.
3:2	A1A0	00	Designates the slave address version. These bits will correspond to the two LSB bits.
1:0	VOUT	11	The default output voltage is 0.95V.
<b>Ctrl2, Address 04h, (aka reg4)</b>			
7:6	Reserved		
5	GO	1	Writing to this bit starts a VOUT transition regardless of its initial value.
		0-d	The VOUT is ramped to the default VSEL Value.
4	Discharge	1	When the regulator is disabled, the output voltage is discharged through the SW pin.
		0-d	When the regulator is disabled, the output voltage is not discharged.
3	PGOK (read only)	1	Is high when output is in regulation and V <sub>REF</sub> has stabilized.
		0	Is low during an output voltage transition or when the output is not in regulation.
2:1	SLEW	01-d	V <sub>REF</sub> slews at 0.8mV/μs.
		10	V <sub>REF</sub> slews at 2.2mV/μs.
		11	V <sub>REF</sub> slews at 8.4mV/μs.
0	Reserved		

Note: -d is the default value at startup.

## 2.7V to 5.5V, 6A Constant Frequency Hysteretic Synchronous Buck Regulator with I<sup>2</sup>C Production Data Spec



Dimensional Ref.			
REF.	Min.	Nom.	Max.
A	0.800	0.850	1.000
A1	---	---	0.050
A3	0.203 Ref.		
D	1.950	2.000	2.050
E	2.950	3.000	3.050
D2	0.550	0.600	0.650
E2	0.550	0.600	0.650
---	---		
b1	0.200	0.250	0.300
b2	0.150	0.200	0.250
e	0.500 BSC		
L	0.350	0.400	0.450
L2	0.550	0.600	0.650
Tol. of Form & Position			
aaa	0.10		
bbb	0.10		
ccc	0.10		
ddd	0.05		
eee	0.08		

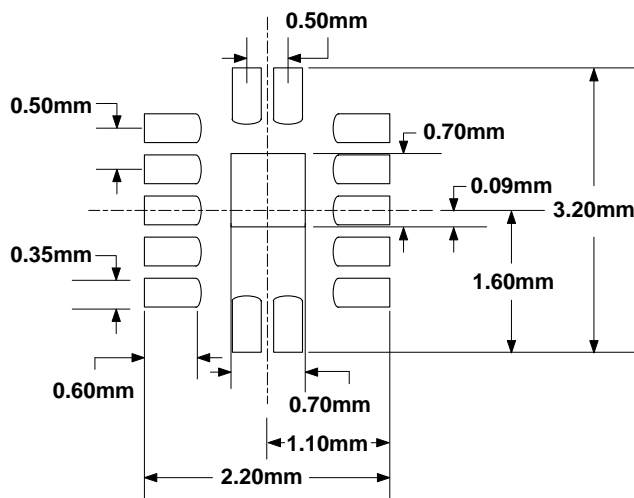
**Notes**  
 1. ALL DIMENSIONS ARE IN MILLIMETERS.  
 2. DIMENSIONING AND TOLERANCING PER JEDEC MO-220.

<b>X</b>	xy	Output Voltage	Slave Address
J	02	0.9V	E0h
K	12	0.9V	E2h
L	22	0.9V	E4h
M	32	0.9V	E6h
N	03	0.95V	E0h
P	13	0.95V	E2h
R	23	0.95V	E4h
S	33	0.95V	E6h

### Package Dimensions



## Land Pattern Recommendation



### LQ 14-Pin VQFN Package Land Pattern

**Disclaimer:**

This PCB land pattern recommendation is based on information available to Microsemi by its suppliers. The actual land pattern to be used could be different depending on the materials and processes used in the PCB assembly, end user must account for this in their final layout. Microsemi makes no warranty or representation of performance based on this recommended land pattern.

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