Power Matters.[™]

Microsemi.

Improving the Performance of your DC-DC Forward Converter using I²MOSTM MOSFET Technology

Microsemi Space Forum 2015

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1icrosemi

PACE FORUM

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I²MOS advantages

- Highest available SEE performance, 85-90 MeV at full rated BVDss
- Highest Avalanche capability: 5X greater than competition
- TID (Total Ionizing Dose) Rating: 100Krad- 500Krad (depending on specific device)
- Commerce Rating: 9A515.e
 - Most Euro countries will not need a license!
- Competitive pricing on new designs







I²MOS FOM versus Competition





SEE results- Microsemi vs. Competitor



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I²MOS[™] MOS P/N Structure

<u>MRH</u>	<u>BVDSS/1</u> <u>0</u>	<u>Channel</u>	<u>ID @ 25C</u>	<u>Package</u>	Screening	RAD LEVEL
	(V)		(A)			
<u>MRH</u>	<u>20</u>	<u>N</u>	<u>22</u>	<u>U3</u>	<u>S</u>	<u>R</u>
Microsemi	20= 200V	Ν		U3= SMD0.5	S= JANS	R= 100K
Rad- Hard	10= 100V	Р		T2= TO- 39	V= JANTXV	F= 300K G= 500K
MOSFET	13= 130V			T3= TO- 257	C= EDU	
	06= 60V			U5= LCC-18		
	03= 30V			C= die		



I²**MOS[™] Planned Products**

Phase 1: I²MOS[™] portfolio, N- Ch, Sz 3

Bvdss (V)	<u>Similar</u> JEDEC Number	<u>Industry</u> Equivalent	<u>MSC p/n</u>	
150	2N7589U3	IRHNJ67134	MRH15N19U3	
200	2N7591U3	IRHNJ67230	MRH20N16U3	
250	2N7593U3	IRHNJ67234	MRH25N15U3	

Phase 2: I²MOS[™] portfolio, N- Ch, Sz 5.5

<u>Bvdss</u> (V)	<u>Similar</u> <u>JEDEC</u> <u>Number</u>	Industry Equivalent	<u>RH2 Base MSC p/n</u>	Package
150	2N7582T1	IRHMS67164	MRH15N45T1	TO-254
150	2N7581U2	IRHNA67164	MRH15N56U1	SMD-2
200	2N7584T1	IRHMS67260	MRH20N45T1	TO-254
200	2N7583U2	IRHNA67260	MRH20N56U1	SMD-2
250	2N7586T1	IRHMS67264	MRH25N45T1	TO-254
250	2N7585U2	IRHNA67264	MRH25N56U1	SMD-2





TO-254





Efficiency Performance of I²MOS

DC-DC Forward Converter (Resonant Reset Topology) MRH25N15U3 vs. IRHNJ67234



DC-DC Criteria

- Create a Circuit to Reveal Differences in MOSFET Power Losses
- Parallel-Inductor Isolated Forward DC-DC Converter
- Improved Efficiency
 - Resonant Transformer Reset
 - Lower DC Losses in Inductors and Schottky Rectifiers
- Use Vdd = 50Vdc
 - Peak of Resonant Reset Voltage Will Be: V_{dd} + ($I_d(pk) * L_m / C_r$)
 - L_m is Power Transformer Magnetizing Inductance (~120uH)
 - $-C_r$ is the Resonance Capacitance = $C_{oss} \parallel C_j (N_s/N_p) * (~810 pF)$
 - Worst Case Resonance Peak at f_{sw} = 350kHz and V_{out} = 5.0Vdc (~120Vpk)



Efficiency Test Criteria

- Voltage De- Rating= 50%
 - Use Microsemi MRH25N15U3 and IR IRHNJ67234, 250V devices.
 (200V Device May Be Used For Higher Efficiency With Lower Voltage Margin.)
- 100W Maximum Output Power
 - 20A Maximum Output Current
 - -66W For Vout = 3.3Vdc
 - -100W For Vout = 5.0Vdc





Efficiency Test Circuit Schematic





Efficiency Test Circuit Features

- Circuit Used For Power Switch Efficiency Comparison:
 - Circuit Uses U3 Packages For All Power Functions
 - Small Size
 - Ease of Thermal Management
 - Output Uses Paralleled Output Stage For Increased Efficiency
 - Schottkies and Inductors Share Current ~50:50
 - DC Power Losses Reduced by ~1/4 1/3!
 - Optimized for 3.3Vdc < V_{out} < 5Vdc
 - Optimized For 1A < I_{out} < 20A
 - Optimized For 350kHz < f_{sw} < 500kHz
 - Uses COTS Micrel MIC4424 Gate Driver IC
 - Rad-Hard Equivalents Available from Intersil



Efficiency Test Parameters

- DC Output (Vout) Set By Varying Input Duty Cycle
 - Duty Cycle = Desired $V_{out} * (N_s/N_p) / V_{dd}$
- Efficiency:

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$$\begin{split} \eta &= P_{out} / \left(P_{in} + P_{bias} \right) = \left(V_{out} * I_{out} \right) / \left(\left(V_{dd} * I_{dd} \right) + \left(V_{bias} * I_{bias} \right) \right) \\ I_{out} &= Set, Varied from 1A to 20A \\ V_{dd} &= Set, Constant = 50Vdc \\ V_{bias} &= Set, Constant = 12Vdc \\ V_{out} is Set By Varying the Input Duty Cycle \\ I_{dd} and I_{bias} Are Measured at Each Operating Point \end{split}$$



MOSFET Losses

- Key Contributors to Power MOSFET Switch Losses:
 - DC Losses: I_d(rms)² * R_{ds}(on) * D
 - AC Losses: Gate + Switching
 - –Gate Input Losses: $Q_{gt} * V_{bias} * f_{sw}$
 - -Drain Switching Losses: ~ $V_{dd} * I_d(rms) * (t_r + t_f) * fsw / 2)$ + ($C_{oss} * V_{dd}^2 * f_{sw}$)







Forward Converter Design Parameters

- Duty Cycle = D = $(V_{out} / V_{in}) * (N_s / N_p) = t_{on} / t_{off}$
- $I_d(pk) = (I_d(avg) / D) + (0.5 * (V_{dd} * t_{on}) / L_m)$
- $V_{res}(pk) = V_{dd} \Box + I_d(pk) * (L_m / C_r)^{0.5}$
- $t_{res} = \pi * (L_m * C_r)^{0.5}$
- $C_r = C_{oss} + (C_j / (N_p/N_s))$
 - C_j is the Output Schottky Junction Capacitance



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Measured Drain-Source Voltages





$$V_{DS}(pk) = 122V$$

 $V_{DS}(pk) = 148V$

 $V_{out} = 5.0$ Vdc, $I_{out} = 20$ Adc, $f_{sw} = 350$ kHz



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Efficiency Parameters

- From Data Sheet Parameters:
 - R_{ds}(ON)
 - $-MRH25N15U3 175m\Omega$ max.
 - IRHNJ67234 210m Ω max.
 - IR Device 20% Higher Than Microsemi
 - Q_{gt}
 - -MRH25N15U3 32nC typ. (est. 40nC max.)
 - IRHNJ67234 50nC max. (est. 40nC typ.)
 - Coss
 - MRH25N15U3 275pF typ.
 - IRHNJ67234 187pF typ.



Efficiency Data- +3.3Vout, 350 Khz.



At higher currents the improvement in conduction losses provide an advantage.



Efficiency Data, Vout = 5.0V, 350 Khz.

100W Forward Converter Efficency: $V_{dd} = 50Vdc$, $V_{out} = 5.0Vdc$, $f_{sw} = 350kHz$



At higher currents the improvement in conduction losses is slightly better at 5.0Vout vs. 3.3Vout.



Efficiency Data, +3.3Vout, 500 Khz.

100W Forward Converter Efficency: $V_{dd} = 50Vdc$, $V_{out} = 3.3Vdc$, $f_{sw} = 500kHz$



At 500 Khz. There are more switiching losses in both parts but I²MOS part maintains the advantage.



Efficiency Data, 5.0Vout, 500 Khz.

100W Forward Converter Efficency: $V_{dd} = 50Vdc$, $V_{out} = 5.0Vdc$, $f_{sw} = 500kHz$



Efficiency improvements @ higher currents when Vout = 5.0V





Avalanche Energy Performance

MRH25N15U3 vs. IRHNJ67234



Avalanche Basics

- Avalanche Performance Indicates Ruggedness of MOSFET
 - Energy Handling Capability
 - Repetitive
 - Single Pulse
 - Specified in Joules (V * I * t)
- "Unconstrained" Inductors Cause Excursions to V_{BR}(DSS)
 - Energy ~ (L * I_d(pk)² / 2) * (1 (V_{dd} / V_{BR}(DSS))
 - Junction Dissipates Enormous Instantaneous Power
 - If $V_{BR}(DSS) = 250V$ and $I_d(pk)$ = 10A, $P_{inst} = 2500W!$
- The Greater the Avalanche Energy Rating, The Better





Data sheet Specs & Avalanche Test Circuit





Avalanche Test PCB



Board Size = 5.8 x 4.1 x 0.063", 4 Layer FR-4, Double Sided



Avalanche Test Procedure



Ideal Waveforms



Measured Avalanche Performance

MRH25N15U3, 7.5mJ

MRH25N15U3, 15mJ



 $V_{BR}(DSS) = 283V, I_d(pk) = 12A, t_{on} = 29.5us$

 $V_{BR}(DSS) = 304V, I_d(pk) = 11A, t_{on} = 71.5us$



Measured Avalanche Performance

MRH25N15U3, 300mJ



 $V_{BR}(DSS) = 314V$, $I_d(pk) = 15A$, $t_{on} = 750us$

MRH25N15U3, 300mJ (Expanded)



 $V_{BR}(DSS) = 314V$, $I_d(pk) = 15A$, $t_{on} = 750us$



Measured Avalanche Performance

IRHNJ67234, 7.5mJ

IRHNJ67234, 56mJ





40.0µs

→▼120.400ι



25.0MS/s 10k points 4) J 2.28 V

WWWWWWWW

Summary

Efficiency

- MRH25N15U3 Demonstrated More Efficient Than IR IRHNJ67234
 - By up to 2.75%
- MGN25N15U3 Efficiency holds up over the full current range of 5A – 20A. Especially at higher load currents
- IRHNJ67234 Efficiency decreases due to higher conduction Losses
 - Useful Output Current Range Must Be De-Rated to 18A
- Increased Losses Mean More Aggressive Thermal Management Required (Bigger Heat Sink for Lower θ_{JA})

Avalanche Capability

- Microsemi MRH25N15U3
 Demonstrated 2X Repetitive Avalanche Capability Over IR IRHNJ67234
- Microsemi MRH25N15U3
 Demonstrated 5.4X Single Event Avalanche Capability Over IR IRHNJ67234





Thank You



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